

ECS455: Chapter 5 OFDM

5.4 Cyclic Prefix (CP)



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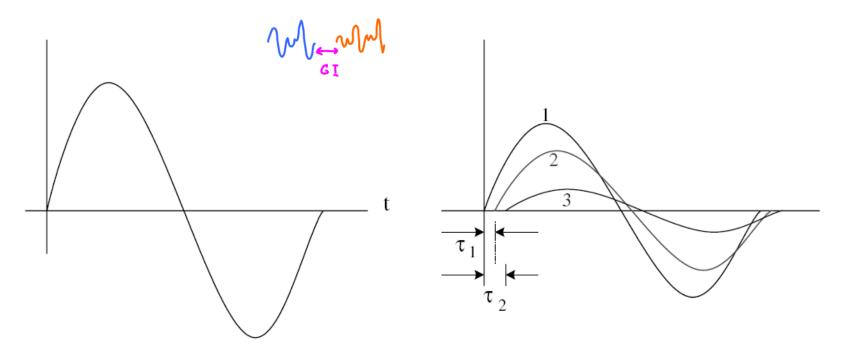
BKD 3601-7 Wed 9:20-11:20

Three steps towards modern OFDM

- 1. Mitigate Multipath (ISI): Decrease the rate of the original data stream via multicarrier modulation (FDM)
- 2. Gain Spectral Efficiency: Utilize orthogonality
- 3. Achieve Efficient Implementation: FFT and IFFT
- Extra step: Completely eliminate ISI and ICI
 - Cyclic prefix

Cyclic Prefix: Motivation (1)

• Recall: Multipath Fading and Delay Spread



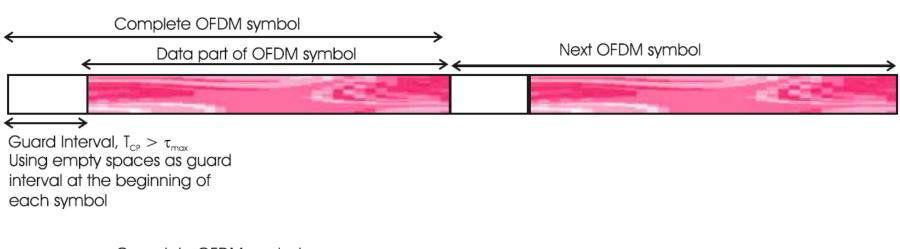
Transmitted Signal

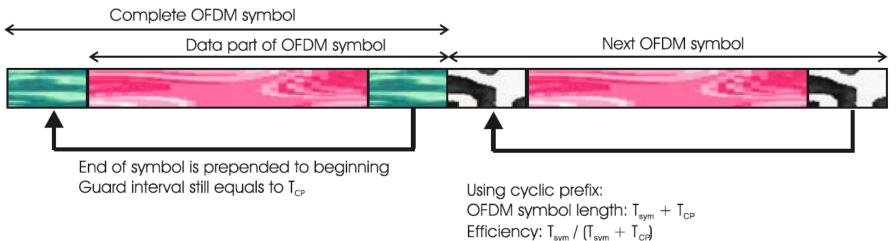
Received Signal

Cyclic Prefix: Motivation (2)

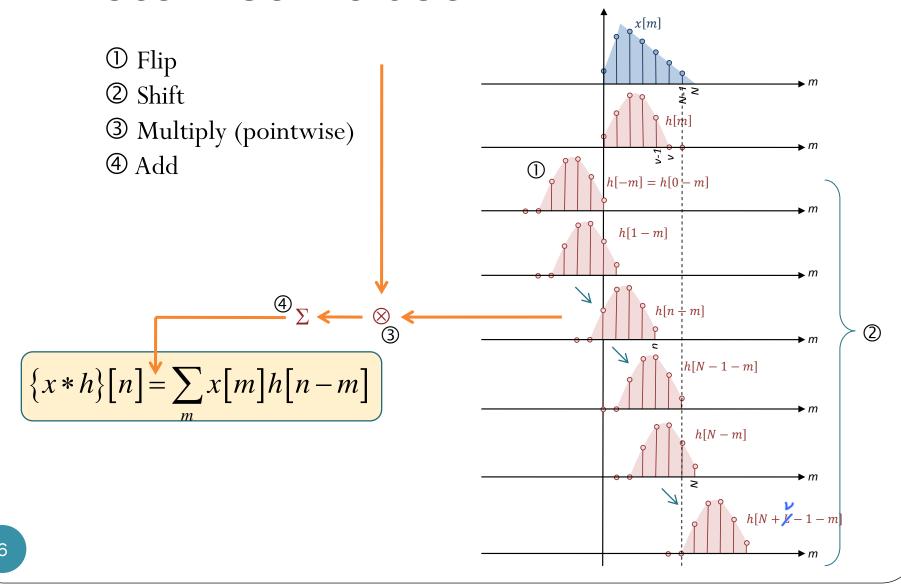
- OFDM uses large symbol duration T_s
 - compared to the duration of the impulse response τ_{max} of the channel
 - to reduce the amount of ISI
- Q: Can we "eliminate" the multipath (ISI) problem?
- A: To reduce the ISI, add **guard interval** larger than that of the estimated delay spread.
- If the guard interval is left empty, the orthogonality of the sub-carriers no longer holds, i.e., ICI (inter-channel interference) still exists.
- Solution: To prevent both the ISI as well as the ICI, OFDM symbol is cyclically extended into the guard interval.

Cyclic Prefix



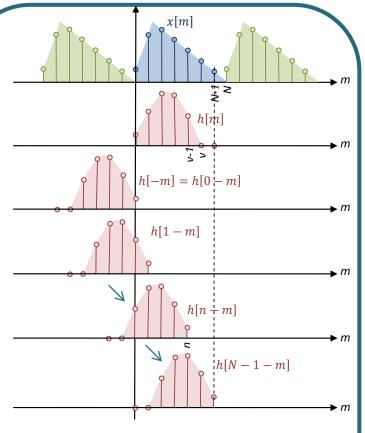


Recall: Convolution

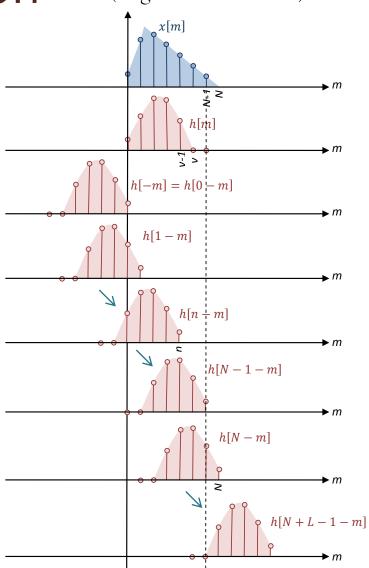


Circular Convolution

(Regular Convolution)



Replicate x (now it looks periodic) Then, perform the usual convolution only on n = 0 to N-1



Circular Convolution: Examples 1

```
Find

resular convolution

[1 \ 2 \ 3]*[4 \ 5 \ 6]

= [+ 13 28 27 18]
```

```
[1 \ 2 \ 3] * [4 \ 5 \ 6] \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \ ^{1} \ ^{2} \ ^{3} \
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$$[1 \ 2 \ 3 \ 0 \ 0] \circledast [4 \ 5 \ 6 \ 0 \ 0] = [4 \ 13 \ 28 \ 27 \ 18]$$

Discussion

- Regular convolution of an N_1 —point vector and an N_2 —point vector gives (N_1+N_2-1)-point vector.
- *Circular convolution* is performed between two equallength vectors. The results also has the same length.
- Circular convolution can be used to find the regular convolution by **zero-padding**.
 - Zero-pad the vectors so that their length is N_1+N_2-1 .
 - Example:

$$\begin{bmatrix} 1 & 2 & 3 & 0 & 0 \end{bmatrix} \circledast \begin{bmatrix} 4 & 5 & 6 & 0 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 2 & 3 \end{bmatrix} * \begin{bmatrix} 4 & 5 & 6 \end{bmatrix}$$

• In modern OFDM, we want to perform circular convolution via regular convolution.

Circular Convolution in Communication

- We want the receiver to obtain the circular convolution of the signal (channel input) and the channel.
- Q: Why?
- A:
 - CTFT: convolution in time domain corresponds to multiplication in frequency domain.
 - This fact does not hold for DFT.
 - **DFT**: circular **convolution** in (discrete) time domain corresponds to **multiplication** in (discrete) frequency domain.
 - We want to have multiplication in frequency domain.
 - So, we want circular convolution and not the regular convolution.
- Problem: Real channel does regular convolution.
- Solution: With **cyclic prefix**, regular convolution can be used to create circular convolution.

0 1 2 3

 $\begin{bmatrix} 1 & -2 & 3 & 1 & 2 \end{bmatrix} \circledast \begin{bmatrix} 3 & 2 & 1 & 0 & 0 \end{bmatrix} = ?$

Solution:

Let's look closer at how we carry out the circular convolution operation.

Recall that we replicate the *x* and then perform the regular convolution (for *N* points)

$$1 \times 1 + 2 \times 2 + 1 \times 3 = 1 + 4 + 3 = 8$$

 $2 \times 1 + 1 \times 2 + (-2) \times 3 = 2 + 2 - 6 = -2$

$$1 \times 1 + (-2) \times 2 + 3 \times 3 = 1 - 4 + 9 = 6$$

$$(-2) \times 1 + 3 \times 2 + 1 \times 3 = -2 + 6 + 3 = 7$$

$$3 \times 1 + 1 \times 2 + 2 \times 3 = 3 + 2 + 6 = 11$$

$$\begin{bmatrix} 1 & -2 & 3 & 1 & 2 \end{bmatrix} \circledast \begin{bmatrix} 3 & 2 & 1 & 0 & 0 \end{bmatrix} = \begin{bmatrix} 8 & -2 & 6 & 7 & 11 \end{bmatrix}$$

Goal: Get these numbers using regular convolution

 $[1 \quad -2 \quad 3 \quad 1 \quad 2] \circledast [3 \quad 2 \quad 1 \quad 0 \quad 0] = ?$

Observation: We don't need to replicate the *x* indefinitely. Furthermore, when *h* is shorter than *x*, we need only a part of one replica.

$$1 \times 1 + 2 \times 2 + 1 \times 3 = 1 + 4 + 3 = 8$$

$$2 \times 1 + 1 \times 2 + (-2) \times 3 = 2 + 2 - 6 = -2$$

Not needed in the calculation

$$1 \times 1 + (-2) \times 2 + 3 \times 3 = 1 - 4 + 9 = 6$$

$$(-2)\times 1 + 3\times 2 + 1\times 3 = -2 + 6 + 3 = 7$$

$$3 \times 1 + 1 \times 2 + 2 \times 3 = 3 + 2 + 6 = 11$$

$$\begin{bmatrix} 1 & -2 & 3 & 1 & 2 \end{bmatrix} \circledast \begin{bmatrix} 3 & 2 & 1 & 0 & 0 \end{bmatrix} = \begin{bmatrix} 8 & -2 & 6 & 7 & 11 \end{bmatrix}$$

Try this: use only the necessary part of the replica and then convolute (regular convolution) with the channel.

$$\begin{bmatrix} 1 & 2 & 1 & -2 & 3 & 1 & 2 \end{bmatrix} * \begin{bmatrix} 3 & 2 & 1 \end{bmatrix} = ?$$

Copy the last v samples of the symbols at the beginning of the symbol.

This partial replica is called the **cyclic prefix**.

We now know that

• Similarly, you may check that

$$\begin{bmatrix} -2 & 1 & 2 & 1 & -3 & -2 & 1 \end{bmatrix} * \begin{bmatrix} 3 & 2 & 1 \end{bmatrix} = \begin{bmatrix} -6 & -1 & 6 & 8 & -5 & -11 & -4 & 0 & 1 \end{bmatrix}$$
Cyclic Prefix
$$\begin{bmatrix} 2 & 1 & -3 & -2 & 1 \end{bmatrix} * \begin{bmatrix} 3 & 2 & 1 & 0 & 0 \end{bmatrix}$$

• We know, from Example 2, that

```
[ 1 2 1 -2 3 1 2] * [3 2 1] = [ 3 8 8 -2 6 7 11 5 2]

And that

[-2 1 2 1 -3 -2 1] * [3 2 1] = [-6 -1 6 8 -5 -11 -4 0 1]
```

Check that

```
[ 1 2 1 -2 3 1 2 0 0 0 0 0 0 0] * [3 2 1]

= [ 3 8 8 -2 6 7 11 5 2 0 0 0 0 0 0 0]

and

[ 0 0 0 0 0 0 0 -2 1 2 1 -3 -2 1] * [3 2 1]

= [ 0 0 0 0 0 0 0 -6 -1 6 8 -5 -11 -4 0 1]
```

We know that

```
[ 1 2 1 -2 3 1 2] * [3 2 1] = [ 3 8 8 -2 6 7 11 5 2]
[-2 1 2 1 -3 -2 1] * [3 2 1] = [-6 -1 6 8 -5 -11 -4 0 1]
```

• Using Example 3, we have

```
 \begin{bmatrix} 1 & 2 & 1 & -2 & 3 & 1 & 2 & -2 & 1 & 2 & 1 & -3 & -2 & 1 \end{bmatrix} * \begin{bmatrix} 3 & 2 & 1 \end{bmatrix} 
 = \begin{bmatrix} \begin{bmatrix} 1 & 2 & 1 & -2 & 3 & 1 & 2 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ + \begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 & 0 & -2 & 1 & 2 & 1 & -3 & -2 & 1 \end{bmatrix} \end{bmatrix} * \begin{bmatrix} 3 & 2 & 1 \end{bmatrix} 
 = \begin{bmatrix} 3 & 8 & 8 & -2 & 6 & 7 & 11 & 5 & 2 & 0 & 0 & 0 & 0 & 0 & 0 \\ + \begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 & -6 & -1 & 6 & 8 & -5 & -11 & -4 & 0 & 1 \end{bmatrix} 
 = \begin{bmatrix} 3 & 8 & 8 & -2 & 6 & 7 & 11 & -1 & 1 & 6 & 8 & -5 & -11 & -4 & 0 & 1 \end{bmatrix}
```

Putting results together...

- Suppose $\underline{x}^{(1)} = [1, -2, 3, 1, 2]$ and $\underline{x}^{(2)} = [2, 1, -3, -2, 1]$
- Suppose h = [3 2 1]
- At the receiver, we want to get
 - $[1 -2 \ 3 \ 1 \ 2]$ \star $[3 \ 2 \ 1 \ 0 \ 0] = [8 -2 \ 6 \ 7 \ 11]$
 - [2 1 -3 -2 1] * [3 2 1 0 0] = [6 8 -5 -11 -4]
- We transmit [1 2 1 -2 3 1 2 -2 1 2 1 -3 -2 1].

 Cyclic prefix

 Cyclic prefix
- At the receiver, we get

$$\begin{bmatrix} 1 & 2 & 1 & -2 & 3 & 1 & 2 & -2 & 1 & 2 & 1 & -3 & -2 & 1 \end{bmatrix} * \begin{bmatrix} 3 & 2 & 1 \end{bmatrix}$$

$$= \begin{bmatrix} 3 & 8 & 8 & -2 & 6 & 7 & 11 & -1 & 1 & 6 & 8 & -5 & -11 & -4 & 0 & 1 \end{bmatrix}$$

Junk! To be thrown away by the receiver.

Circular Convolution: Key Properties

- Consider an *N*-point signal *x*[*n*]
- Cyclic Prefix (CP) insertion: If x[n] is extended by copying the last V samples of the symbols at the beginning of the symbol:

$$\widehat{x}[n] = \begin{cases} x[n], & 0 \le n \le N - 1 \\ x[n+N], & -v \le n \le -1 \end{cases}$$

• Key Property 1:

$$\{h \circledast x\}[n] = (h * \widehat{x})[n] \text{ for } 0 \le n \le N-1$$

• Key Property 2:

$$\{h \circledast x\}[n] \xrightarrow{\text{FFT}} H_k X_k$$

OFDM with CP for Channel w/ Memory

- We want to send N samples $S_0, S_1, \ldots, S_{N-1}$ across noisy channel with memory.
- First apply IFFT: $S_k \xrightarrow{\text{IFFT}} s[n]$
- Then, add cyclic prefix

$$\widehat{s} = \left\lceil s[N-\nu], \dots, s[N-1], s[0], \dots, s[N-1] \right\rceil$$

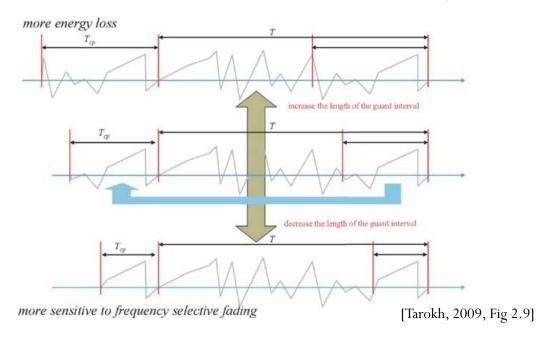
- This is inputted to the channel.
- The output is

$$y[n] = [p[N-v], ..., p[N-1], r[0], ..., r[N-1]]$$

- Remove cyclic prefix to get $r[n] = h[n] \circledast s[n] + w[n]$
- Then apply FFT: $r[n] \xrightarrow{\text{FFT}} R_k$
- By circular convolution property of DFT, $R_k = H_k S_k + W_k$

OFDM System Design: CP

- A good ratio between the CP interval and symbol duration should be found, so that all multipaths are resolved and not significant amount of energy is lost due to CP.
- As a thumb rule, the CP interval must be two to four times larger than the root mean square (RMS) delay spread.



Summary

- The CP at the beginning of each block has two main functions.
- As guard interval, it prevents contamination of a block by ISI from the previous block.
- It makes the received block appear to be periodic of period
 N.
 - Turn regular convolution into circular convolution
 - Point-wise multiplication in the frequency domain

Reference

A. Bahai, B. R. Saltzberg, and M. Ergen, Multi-Carrier Digital
 Communications: Theory and Applications of OFDM, 2nd ed.,
 New York: Springer Verlag, 2004.

